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MAGO Input/Readout Theory

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MAGO 2-Port Circuit

On resonance:

$$\frac{Z_{in}}{Z_c} = \frac{Q_{int}}{Q_e} = \frac{P_e}{P_{int}} \equiv \beta_{in}$$

$$\frac{Z_{out}}{Z_c} = \frac{Q_{int}}{Q_{out}} = \frac{P_t}{P_{int}} \equiv \beta_{out}$$

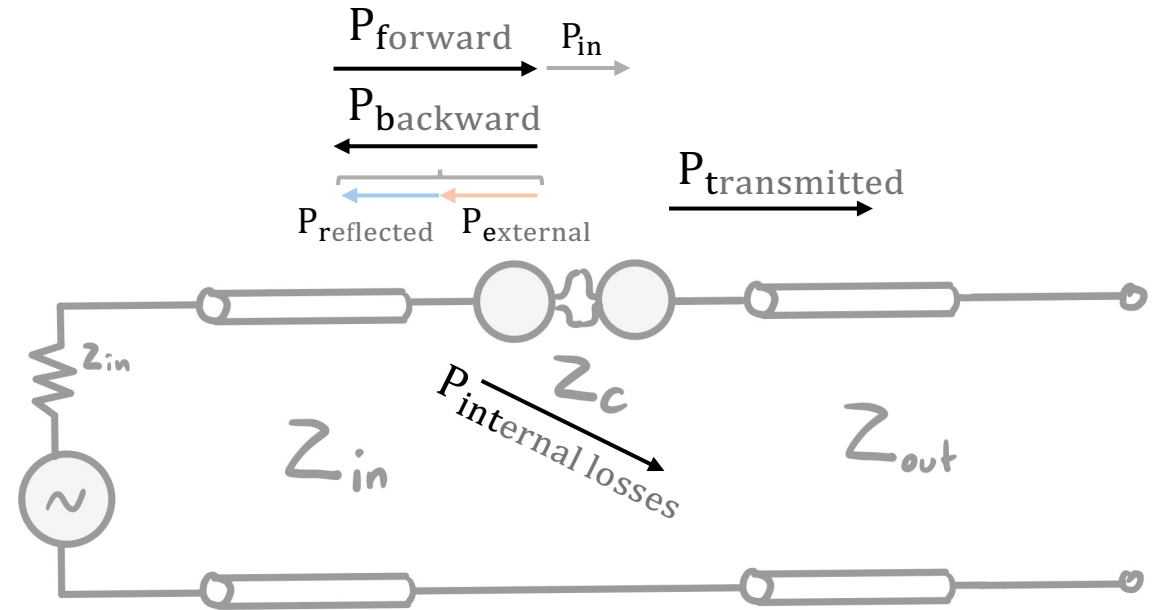
In steady state: $P_{in} = P_e + P_{int} + P_t$

$$\frac{P_b}{P_f} = \left(\frac{1 + \beta_{out} - \beta_{in}}{1 + \beta_{in} + \beta_{out}} \right)^2, \quad \frac{P_t}{P_f} = \frac{4 \beta_{in} \beta_{out}}{(1 + \beta_{in} + \beta_{out})^2}, \quad \frac{P_{int}}{P_f} = \frac{4 \beta_{in}}{(1 + \beta_{in} + \beta_{out})^2}, \quad \text{Stored energy } U_0 = \frac{4 \beta_{in}}{(1 + \beta_{in} + \beta_{out})^2} \frac{Q_{int}}{\omega_0} P_f$$

No backwards power, maximal stored energy, maximal transmission for $\beta_{in} = \beta_{out} + 1$ (if β_{out} is fixed)

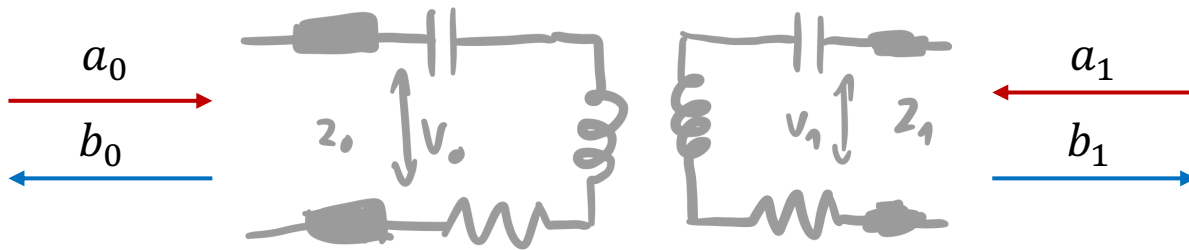
‘critical coupling’ $\frac{P_b}{P_f} = 0, \quad \frac{P_t}{P_f} = \frac{\beta_{out}}{1 + \beta_{out}}, \quad \frac{P_{int}}{P_f} = \frac{1}{1 + \beta_{out}}, \quad \text{Stored energy } U_0 = \frac{1}{1 + \beta_{out}} \frac{Q_{int}}{\omega_0} P_f, \quad Q_L = \frac{1}{1 + \beta_{out}} \frac{Q_{int}}{2}$

$\beta_{in} = 1$ is not really critical coupling anymore!



Taking cavity coupling into account

Consider forward- (a) and backward- (b) travelling power waves



Impedance of single cavity cell: Z_c
Coupling impedance $Z_{cpl} = i\omega L_{cpl}$

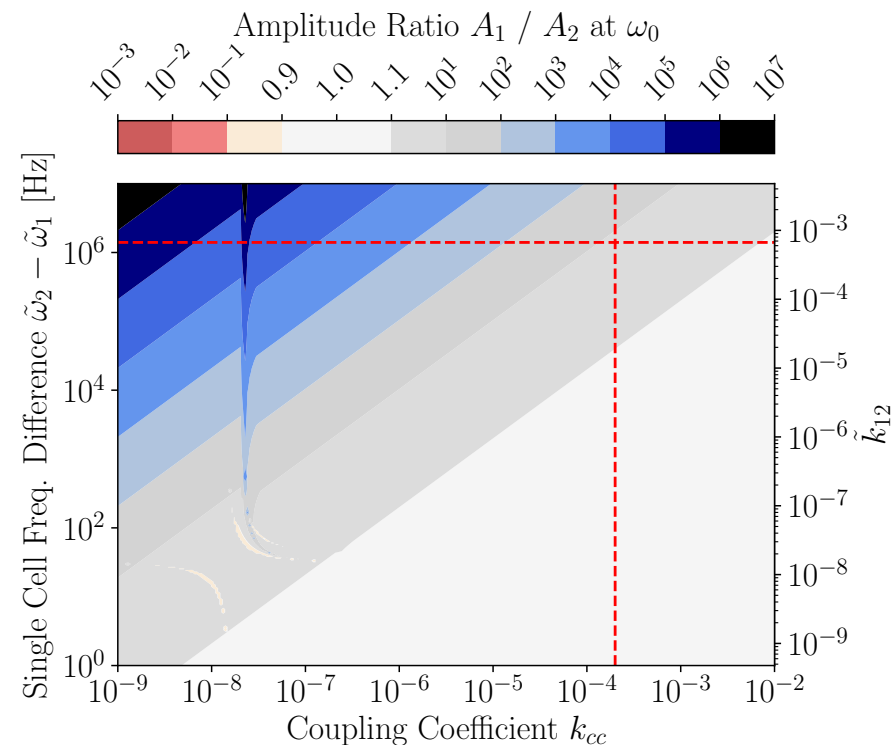
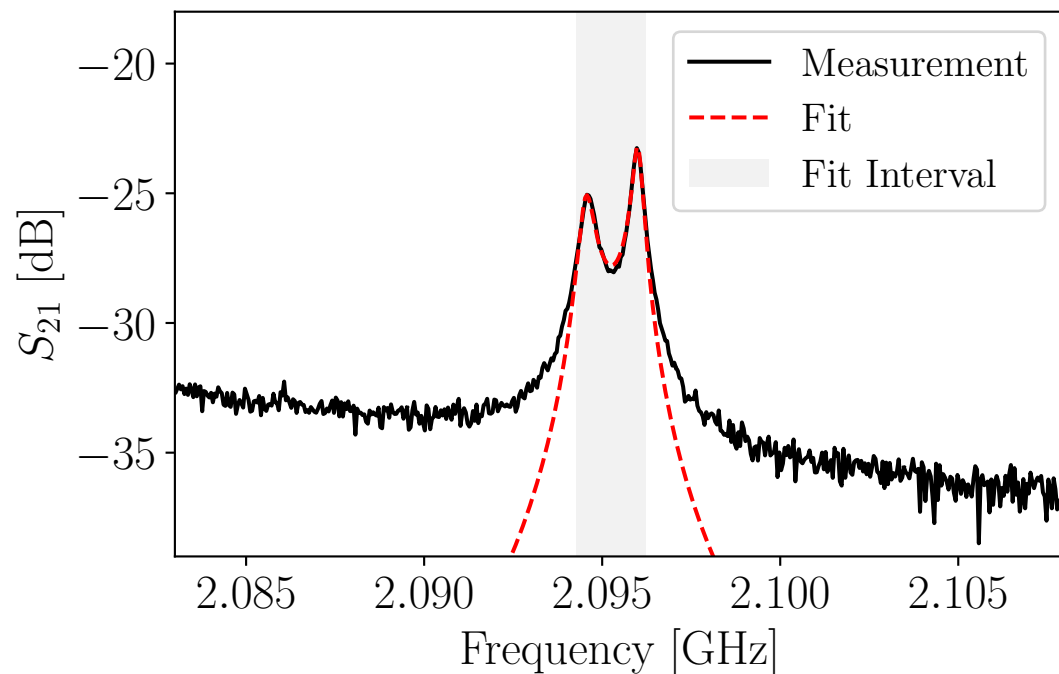
The power relationships can be obtained from the S matrix

$$\begin{pmatrix} b_0 \\ b_1 \end{pmatrix} = S \begin{pmatrix} a_0 \\ a_1 \end{pmatrix}$$

In this case: $S = \frac{1}{(Z_c + Z_1)(Z_c + Z_0) - Z_{cpl}^2} \begin{pmatrix} (Z_c + Z_1)(Z_c - Z_0) - Z_{cpl}^2 & 2\sqrt{Z_0 Z_1} Z_{cpl} \\ 2\sqrt{Z_0 Z_1} Z_{cpl} & (Z_c - Z_1)(Z_c + Z_0) - Z_{cpl}^2 \end{pmatrix}$

If $Z_0 = Z_1$ is chosen, the most transferred power is the value which maximizes S_{21} : $Z_{0,1} = \sqrt{Z_c^2 - Z_{cpl}^2} \approx R_c$ on resonance
 $\Leftrightarrow \beta_{0,1} = 1$ as expected

Fit to real data



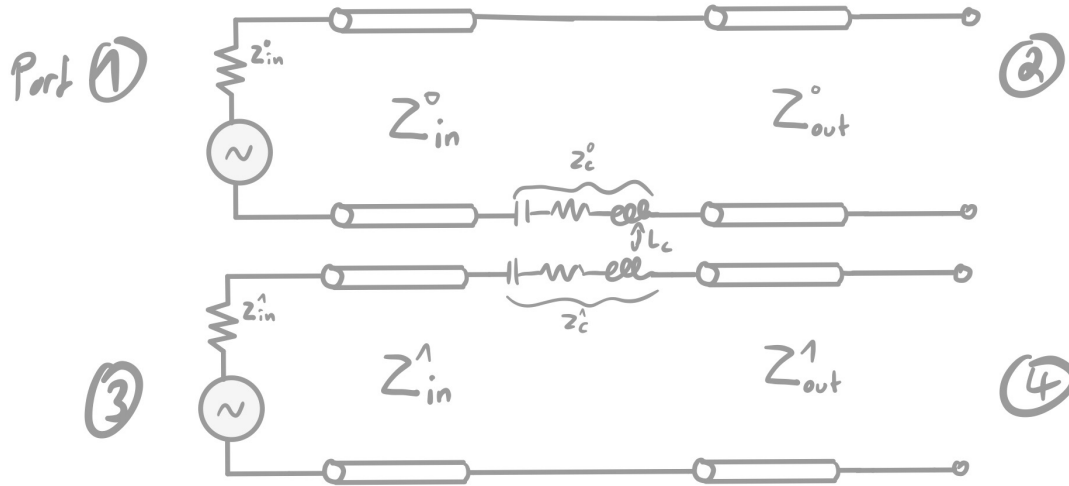
The S_{21} parameter from last slide fit to our RF measurement.

The parameters for the fit curve are $\omega_1 = 2.09457$ GHz, $\omega_2 = 2.09597$ GHz, $k_{cc} = 2 \cdot 10^{-4}$, $Q_1 = 5 \cdot 10^3$ $Q_2 = 8 \cdot 10^3$

These values are robust against the precise choice of the remaining resistances.

E.g. for $R = R_1 = R_2$ any resistances of k Ω to G Ω with $\frac{\sqrt{Z_{in}Z_{out}}}{R} = 0.013$ reproduce the above fit.

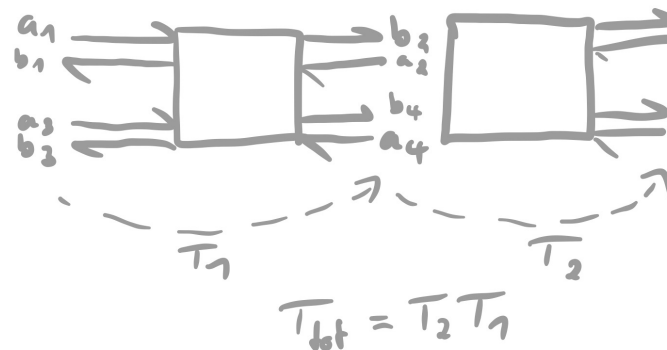
Four port system



$$S_{\text{cavity}} = \frac{1}{Z_{\text{tot}}^2 - Z_{\text{cpl}}^2} \cdot \left[\begin{pmatrix} Z_{\text{tot}}(Z_c - Z_{\text{in}} + Z_{\text{out}}) & 2Z_{\text{tot}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & 2Z_{\text{cpl}}Z_{\text{in}} & -2Z_{\text{cpl}}\sqrt{Z_{\text{in}}Z_{\text{out}}} \\ 2Z_{\text{tot}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & Z_{\text{tot}}(Z_c + Z_{\text{in}} - Z_{\text{out}}) & -2Z_{\text{cpl}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & 2Z_{\text{cpl}}Z_{\text{out}} \\ 2Z_{\text{cpl}}Z_{\text{in}} & -2Z_{\text{cpl}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & Z_{\text{tot}}(Z_c - Z_{\text{in}} + Z_{\text{out}}) & 2Z_{\text{tot}}\sqrt{Z_{\text{in}}Z_{\text{out}}} \\ -2Z_{\text{cpl}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & 2Z_{\text{cpl}}Z_{\text{out}} & 2Z_{\text{tot}}\sqrt{Z_{\text{in}}Z_{\text{out}}} & Z_{\text{tot}}(Z_c + Z_{\text{in}} - Z_{\text{out}}) \end{pmatrix} - Z_{\text{cpl}}^2 \mathbb{1} \right] \quad (5.1)$$

Also want to take magic Tees, and reflections at unused ends into account. To find total S matrix, need to consider transfer matrices T.

$$\begin{pmatrix} b_1 \\ b_2 \\ b_3 \\ b_4 \end{pmatrix} = S \begin{pmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{pmatrix} \quad \text{vs} \quad \begin{pmatrix} b_2 \\ a_2 \\ b_4 \\ a_4 \end{pmatrix} = T \begin{pmatrix} a_1 \\ b_1 \\ a_3 \\ b_3 \end{pmatrix}$$



Finding the S matrix of cascaded system:

- First, find T matrix of all components
- multiply them
- Convert the total T matrix into a S matrix

Cascaded Network

Basic approach



S Matrix

$$\begin{pmatrix} \frac{Z_c + Z_{cpl} - Z_{in} + Z_{out}}{Z_c + Z_{cpl} + Z_{in} + Z_{out}} & \frac{2\sqrt{Z_{in} Z_{out}}}{Z_c + Z_{cpl} + Z_{in} + Z_{out}} & 0 & 0 \\ \frac{2\sqrt{Z_{in} Z_{out}}}{Z_c + Z_{cpl} + Z_{in} + Z_{out}} & \frac{Z_c + Z_{cpl} - Z_{in} - Z_{out}}{Z_c + Z_{cpl} + Z_{in} + Z_{out}} & 0 & 0 \\ 0 & 0 & \frac{Z_c - Z_{cpl} - Z_{in} + Z_{out}}{Z_c - Z_{cpl} + Z_{in} + Z_{out}} & \frac{2\sqrt{Z_{in} Z_{out}}}{Z_c - Z_{cpl} + Z_{in} + Z_{out}} \\ 0 & 0 & \frac{2\sqrt{Z_{in} Z_{out}}}{Z_c - Z_{cpl} + Z_{in} + Z_{out}} & \frac{Z_c - Z_{cpl} - Z_{in} - Z_{out}}{Z_c - Z_{cpl} + Z_{in} + Z_{out}} \end{pmatrix}$$

$$U_0 = \frac{4\beta_{in}}{(1 + \beta_{in} + \beta_{out})^2} \frac{Q_{int}}{\omega_0} P_f \leq \frac{1}{1 + \beta_{out}} \frac{Q_{int}}{\omega_0} P_f$$

for $\beta_{in} = 1 + \beta_{out}$

'Sensitivity enhancement (MAGO)'



$$\begin{pmatrix} \frac{Z_c + Z_{cpl} - Z_{in}}{Z_c + Z_{cpl} + Z_{in}} & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & -\frac{Z_c + Z_{cpl} + Z_{out}}{Z_c - Z_{cpl} + Z_{out}} \end{pmatrix}$$

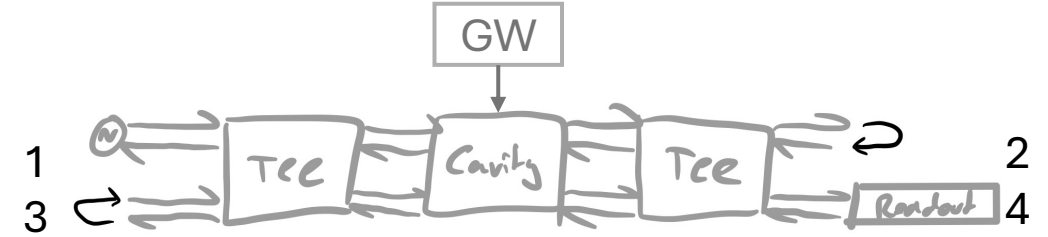
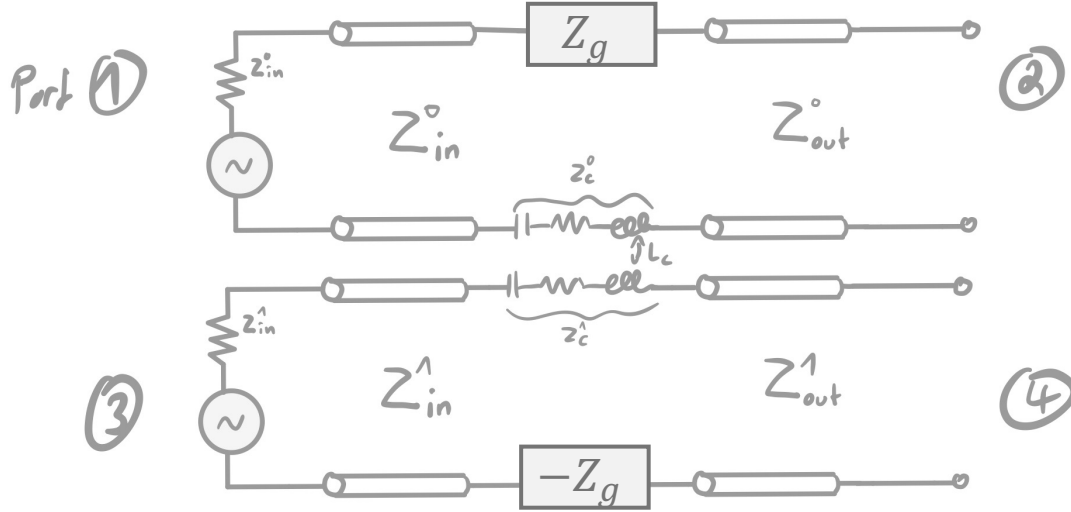
$$U_0 = \frac{4\beta_{in}}{(1 + \beta_{in})^2} \frac{Q_{int}}{\omega_0} P_f \leq \frac{Q_{int}}{\omega_0} P_f$$

for $\beta_{in} = 1$ and $\beta_{out} \in \mathbb{R}$

- Can couple to both modes **separately!**
- Pump mode can be driven at ideal efficiency

(assuming negligible power is lost in the reflection and antennas)

Consider GW Signal



Add small anti-symmetric excitation through positive and negative impedance in both circuits

$$(Z_{in} + Z_{out} + Z_c)I_1 = -Z_g I_1 - Z_{cpl} I_2 + U_0$$

$$(Z_{in} + Z_{out} + Z_c)I_2 = +Z_g I_2 - Z_{cpl} I_1 + U_0$$

→ Signal strength is proportional to pump energy

→ But at the same frequency as pump mode ($\omega_{GW} = 0$)

Basic approach $\left(\frac{P_{sig}}{P_f}\right)^{1/2} = \frac{2Z_g \sqrt{Z_{in} Z_{out}}}{(Z_c + Z_{cpl} + Z_{in} + Z_{out})(Z_c - Z_{cpl} + Z_{in} + Z_{out})}$

‘Sensitivity enhancement (MAGO)’ $\left(\frac{P_{sig}}{P_f}\right)^{1/2} = \frac{2Z_g \sqrt{Z_{in} Z_{out}}}{(Z_c + Z_{cpl} + Z_{in})(Z_c - Z_{cpl} + Z_{out})}$

Ratio: $\left(\frac{P_{sig}^{SE}}{P_{sig}^{Basic}}\right)^{1/2} = \frac{(Z_c + Z_{cpl} + Z_{in} + Z_{out})(Z_c - Z_{cpl} + Z_{in} + Z_{out})}{(Z_c + Z_{cpl} + Z_{in})(Z_c - Z_{cpl} + Z_{out})} > 1$ and $\left(\frac{P_{sig}^{SE}}{P_{sig}^{Basic}}\right)^{1/2} \approx \frac{1 + \beta_{out}}{1 + \beta_{in}}$ for $\beta_{out} \gg \beta_{in}$

Summary

- The ‘sensitivity enhancement’ circuit allows **coupling to the pump and signal mode separately** (i.e. overcoupling the *readout* does not affect the loaded Q of the *pump mode*)
 - ⇒ The **pump mode can be loaded with the same energy** using the same forward power irrespective of the output coupling
 - This depends on the quality of the reflection/antennas/magic Tee
- The ‘sensitivity enhancement’ circuit **improves the GW signal power, but not by a lot**
 - Only significantly if $\beta_{out} \gg \beta_{in}$ when more energy can be stored in the pump mode