



Basic Considerations on Spatial Resolution in Patterned Detectors

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Overview

- Warmup:
 - Resolution with binary readout – optimal signal width
- Error of Center-of-Gravity: When do we need a fit?
- Influence of noise on spatial resolution
 - Higher Moments
 - Correlated Noise
 - 2D structures
 - Wide signals
- Error when doing ‘Eta-reconstruction’
 - Search for ‘best’ response function

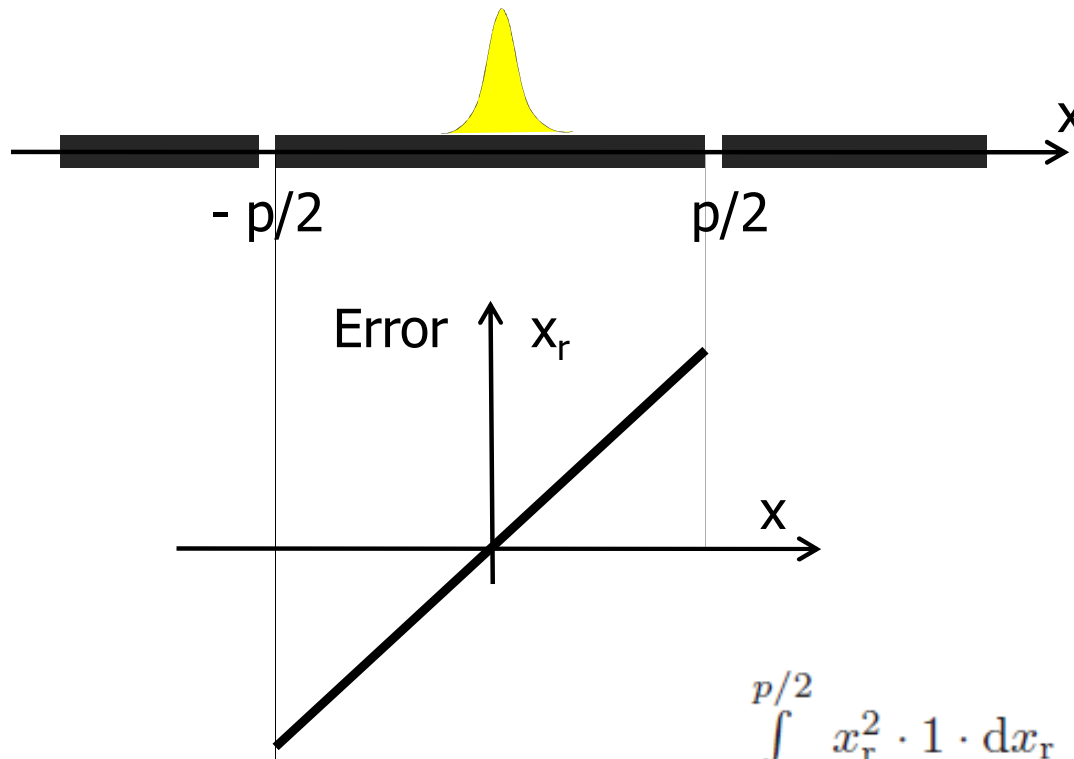


BINARY READOUT OF BOX SIGNALS



Spatial Resolution of Narrow Signals

- Consider very narrow signal
- Only **one** strip is hit
- Reconstructed position = strip center, error = offset in strip



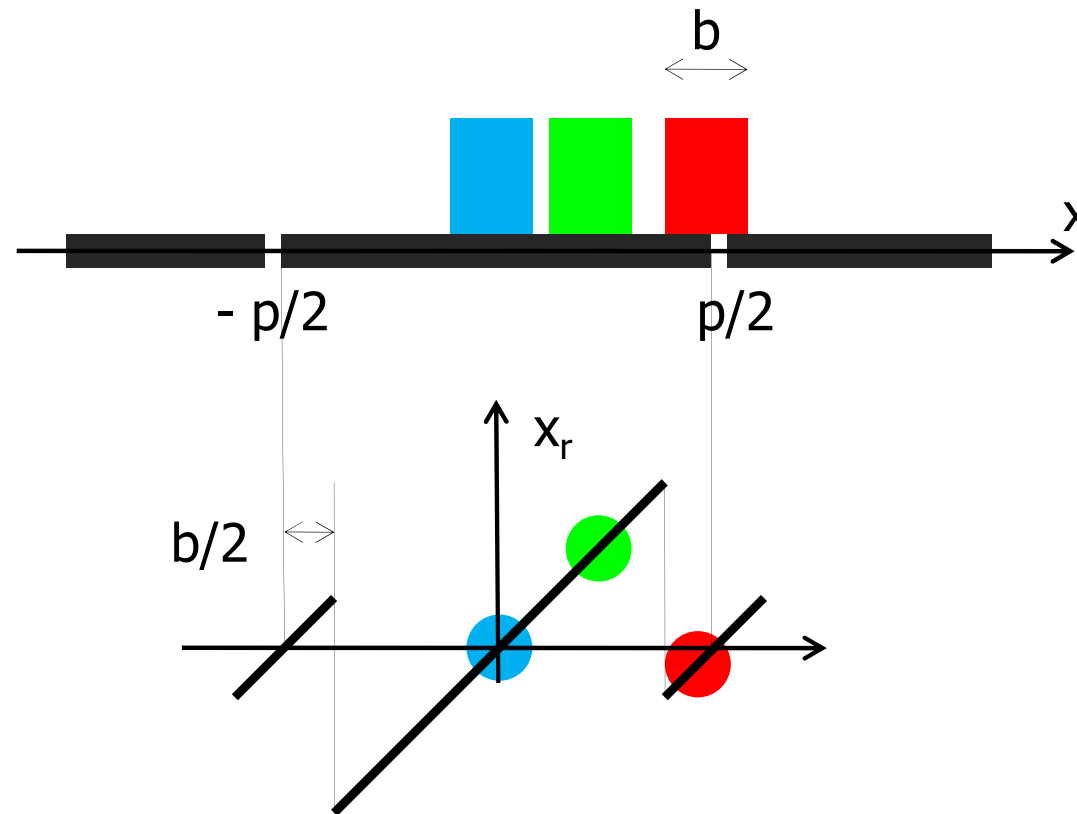
■ Sigma of Error $\sigma_{\text{position}}^2 = \frac{\int_{-p/2}^{p/2} x_r^2 \cdot 1 \cdot dx_r}{\int_{-p/2}^{p/2} 1 \cdot dx_r} = \frac{p^2}{12}$

$$\sigma_{\text{position}} = \frac{p}{\sqrt{12}}$$



Resolution with wider Signals (Binary Readout!)

- Consider 'Box' Signals for simplicity
- When 2 strips are hit \rightarrow reconstruct at edge \rightarrow small error



- Minimum Error for $b = p/2$. Error becomes half: $\sigma = \frac{1}{2} p/\sqrt{12}$
- Note that we have 50% single / 50% double hits

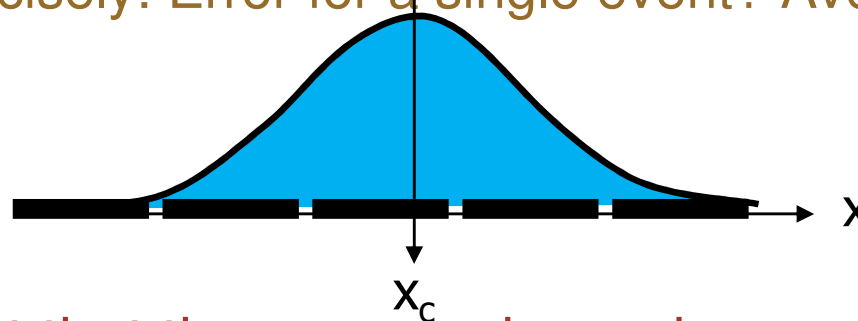


**CENTER-OF-GRAVITY RECONSTRUCTION:
WHEN IS IT SUFFICIENT - OR -
WHEN DO WE NEED A FIT?**



The question:

- A 1D signal with (spatial) shape $f(x)$ falls onto a strip structure with pitch a
- We assume $\int f(x) dx = 1$ and $f(x)$ **symmetric**.
- This generates (analogue) signals on several strips.
- We assume for now that **noise = 0**.
- Question:
What is the reconstruction error for GoG reconstruction ?
 - More precisely: Error for a single event? Average error? Sigma?



- We expect that the answer depends on
 - signal shape
 - Strip pitch a
 - signal position (for single events)

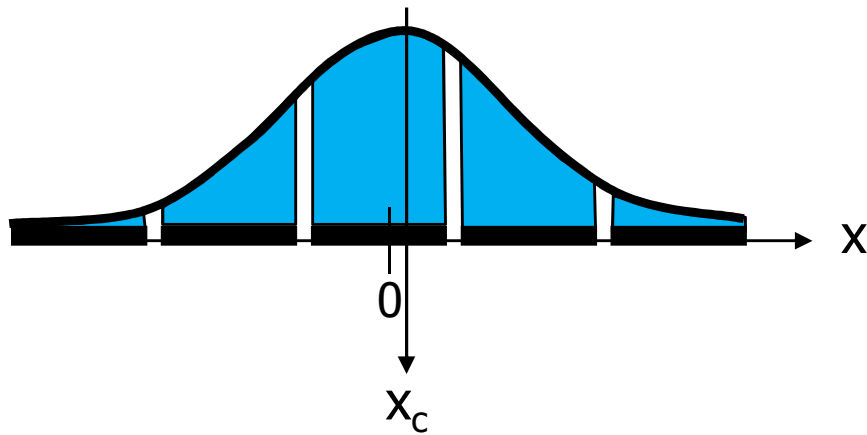


Remark

- The following calculation involves partial integrals over arbitrary function.
 - Normally we must give up soon analytically (consider Gaussians..)
 - But it turns out that we can...
-
- I could only show the result, but I like the fact that so many 'simple' aspects of basic Analysis are required...



1. Signal on Strips



We **assume** the signal on a strip is the integral of $f(x)$

Strip centers are at $m \cdot a$

$$S_m(x_c) = \int_{(m-1/2)a}^{(m+1/2)a} f(x - x_c) dx$$

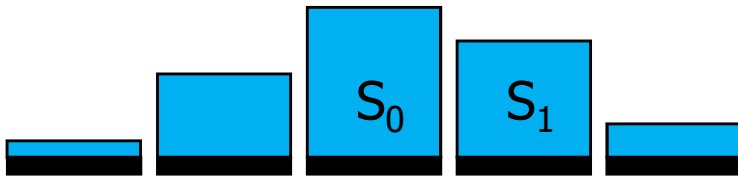
This is the signal in strip m when the charge cloud is centered around x_c

$$= \int_{-\infty}^{\infty} f(x) \cdot \text{Rect}_a(x - (ma - x_c)) dx$$

Box of width a centered at $ma - x_c$



2. Reconstructed Position



$x_{rek} ?$

position reconstructed by CoG
(we assumed a normalized signal)

$$x_{rek}(x_c) = \sum_m m a \cdot S_m(x_c)$$

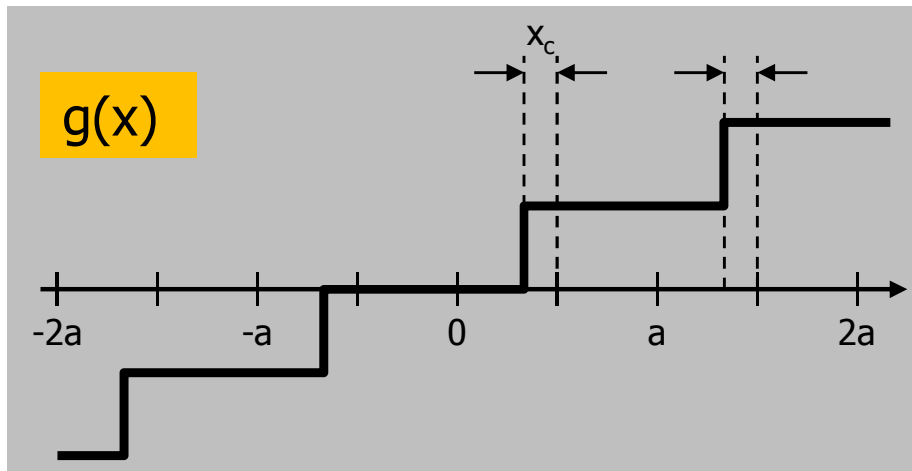
$$= a \int f(x) \underbrace{\sum_m m \cdot \text{Box}_a(x + x_c - m a)}_{\text{Staircase} = g(x)} dx$$

$$= a \int f(x) g(x) dx$$

Staircase = $g(x)$

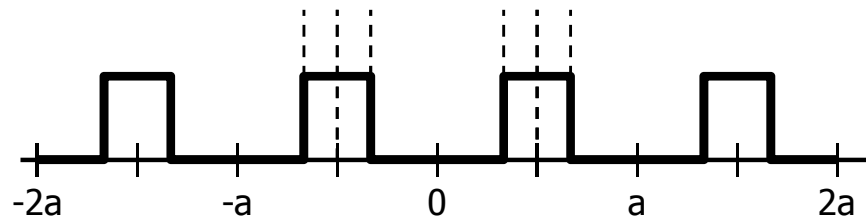
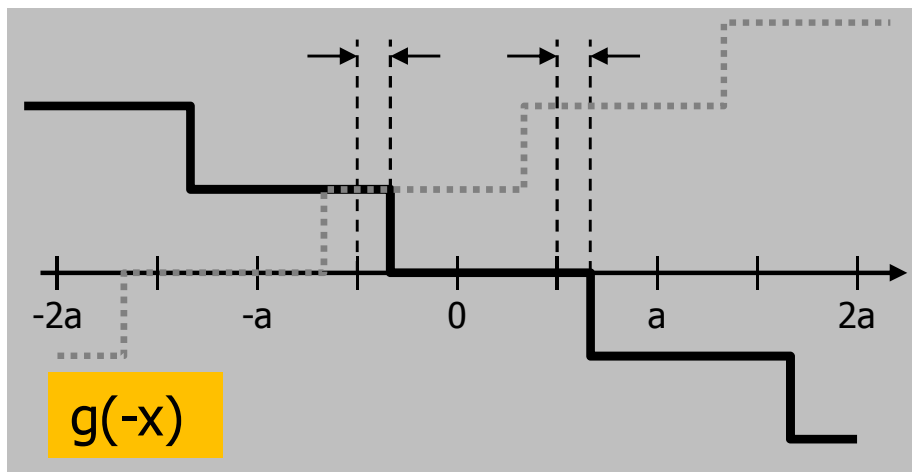


3. Divide Staircase in sym. / antisym. parts



$$g(x) = g_{\text{sym}}(x) + g_{\text{antisym}}(x)$$

$$g_{\text{sym}}(x) = \frac{g(x) + g(-x)}{2}$$



$$g_{\text{sym}}(x) = \frac{1}{2} - \frac{1}{2} \sum_m \text{Box}_{a-2x_c}(x - ma)$$



4. Simplify the integrals, Move to Fourier Space

- Integral of $g_{\text{antisym}}(x)$ is zero because f is symmetric
- We are left with

$$\begin{aligned} x_{rek}(x_c) &= a \int f(x) g_{\text{sym}}(x) dx \\ &= \frac{a}{2} - \frac{a}{2} \int f(x) \sum_m \text{Box}_{a-2x_c}(x - ma) dx \\ &= \frac{a}{2} - \frac{a}{2} \int f(x) \left[\text{Box}_{a-2x_c}(x) \star \text{Comb}_a(x) \right] dx \end{aligned}$$

Write Sum of Boxes
as convolution of a
single Box with
Dirac Comb

- To solve this, move to Fourier Space with

$$\tilde{f}(k) := \int f(x) e^{-2\pi i k x} dx$$

- We can use $\widetilde{a \star b} = \tilde{a} \cdot \tilde{b}$ and $\int a(x) b(x) dx = \int \tilde{a}(k) \tilde{b}(k) dk$
(for symmetrical a, b)



5. Get rid of the Integral

$$x_{rek}(x_c) = \frac{a}{2} - \frac{a}{2} \int f(x) \left[\text{Box}_{a-2x_c}(x) \star \text{Comb}_a(x) \right] dx$$

$$= \frac{a}{2} - \frac{a}{2} \int \tilde{f}(k) \cdot \widetilde{\text{Box}_{a-2x_c}(x)} \cdot \widetilde{\text{Comb}_a(x)} dk$$

This is again a Dirac Comb, i.e. a **sum** of peaks at distances $1/a$

sinc

integral can be carried out. Sum is left

$$= \frac{a}{2} - \frac{a}{2} \sum_{m=-\infty}^{\infty} \tilde{f}\left(\frac{m}{a}\right) \frac{\sin\left(m\pi - \frac{2\pi m x_c}{a}\right)}{m\pi}$$



6. Use Symmetry, Simplify the Sin() function

$$x_{rek} = \frac{a}{2} - \frac{a}{2} \sum_{m=-\infty}^{\infty} \tilde{f}\left(\frac{m}{a}\right) \frac{\sin\left(m\pi - \frac{2\pi m x_c}{a}\right)}{m\pi}$$

Use symmetry
in m.

Treat m=0:
~f(0) = 1,
Sin(m k)/m -> k

$$= x_c - \frac{a}{\pi} \sum_{m=1}^{\infty} \tilde{f}\left(\frac{m}{a}\right) \frac{\sin\left(m\pi - \frac{2\pi m x_c}{a}\right)}{m}$$

Center position
 x_c shows up !

$$\begin{aligned} \sin(m\pi - x) &= \sin(m\pi) \cos(x) - \cos(m\pi) \sin(x) \\ &= -(-1)^m \sin(x) \end{aligned}$$

$$x_{err}(x_c) = \frac{a}{\pi} \sum_{m=1}^{\infty} \frac{(-1)^m}{m} \tilde{f}\left(\frac{m}{a}\right) \sin\left(\frac{2\pi m x_c}{a}\right)$$



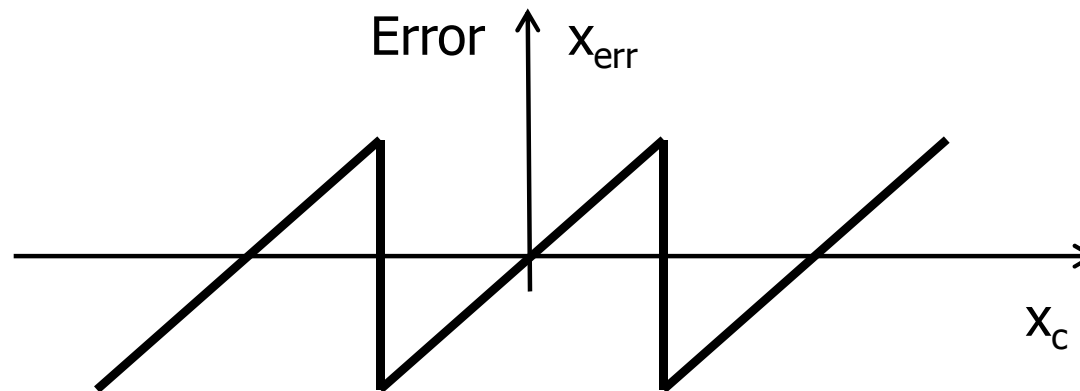
A first check

$$x_{err}(x_c) = \frac{a}{\pi} \sum_{m=1}^{\infty} \frac{(-1)^m}{m} \tilde{f}\left(\frac{m}{a}\right) \sin\left(\frac{2\pi m x_c}{a}\right)$$

- For very narrow $f(x)$, we have $\tilde{f}(k) \rightarrow 1$ so that

$$x_{err}(x_c) = \frac{a}{\pi} \sum_{m=1}^{\infty} \frac{(-1)^m}{m} \sin\left(m\pi \frac{2x_c}{a}\right)$$

- This is the Fourier Series of a Saw-Tooth, as expected!

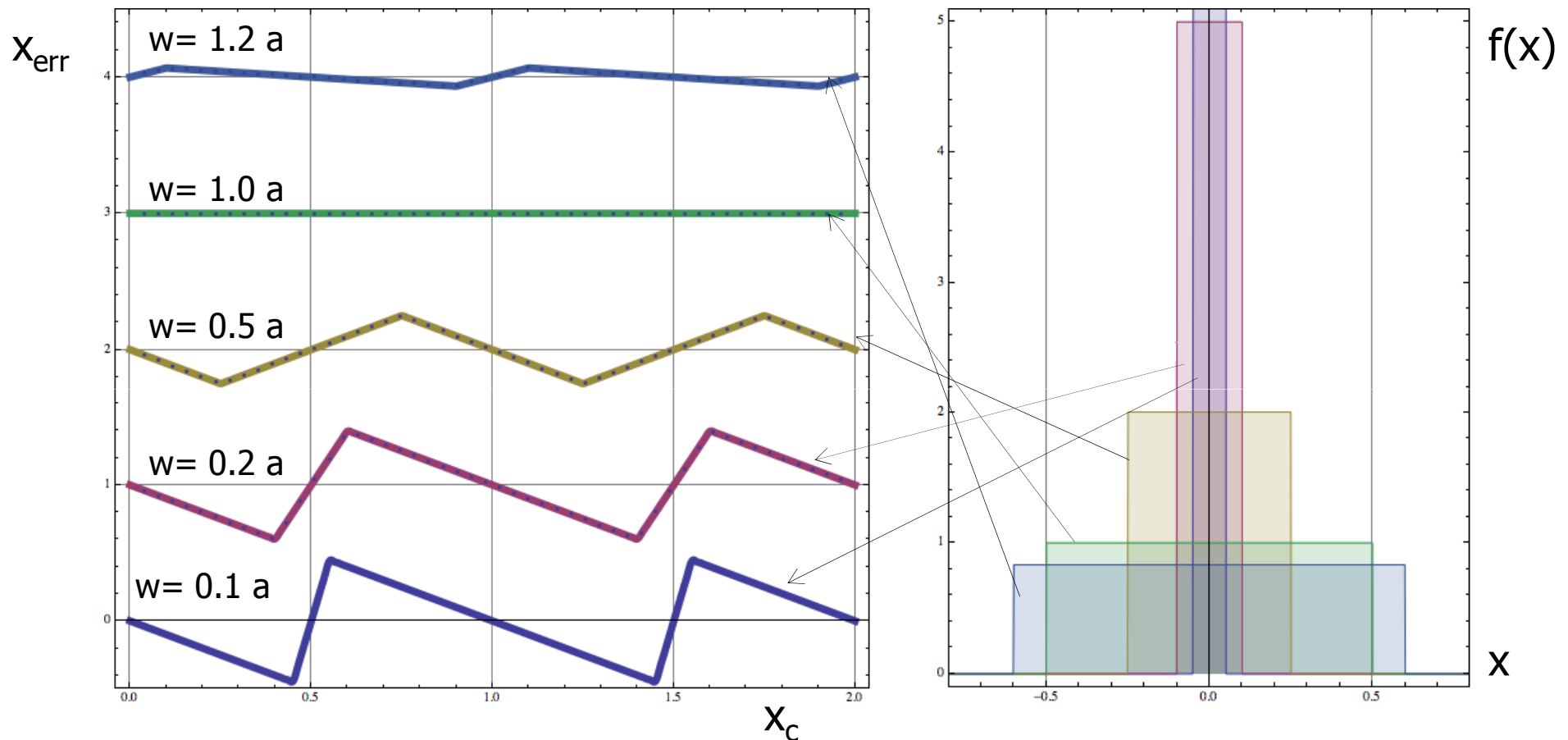




Check with Box $f(x)$

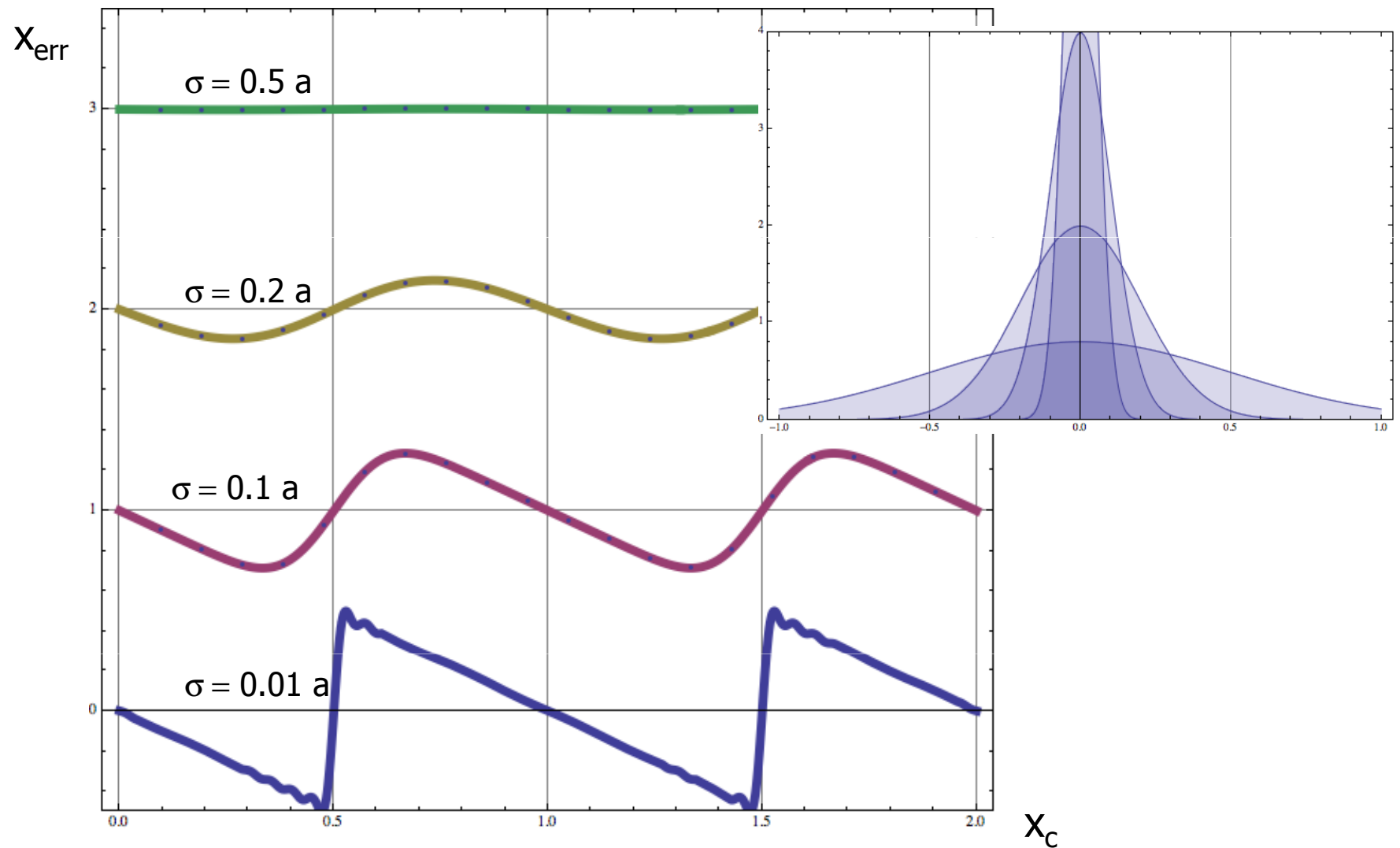
- For Box of width a , $\tilde{f}\left(\frac{m}{a}\right) = \sin(m\pi)/m\pi$ is zero for integer m .
→ reconstruction is perfect..

$$x_{err}(x_c) = \frac{a}{\pi} \sum_{m=1}^{\infty} \frac{(-1)^m}{m} \tilde{f}\left(\frac{m}{a}\right) \sin\left(\frac{2\pi m x_c}{a}\right)$$





Check with Gaussians



- Error already very small for $\sigma = 0.5a$



Going Further: Sigma of x_{err} ?

$$\begin{aligned}\sigma_{\text{rec}}^2 &= \frac{1}{a} \int_{-a/2}^{a/2} x_{\text{err}}^2(x_c) dx_c \\ &= \frac{a}{\pi^2} \sum_{n,m=1}^{\infty} \frac{(-1)^{n+m}}{nm} \tilde{f}\left(\frac{n}{a}\right) \tilde{f}\left(\frac{m}{a}\right) \underbrace{\int_{-a/2}^{a/2} \sin \frac{2\pi n x_c}{a} \sin \frac{2\pi m x_c}{a} dx_c}_{\text{sin() are orthorgonal!}} \\ &= \frac{a}{\pi^2} \sum_{n,m=1}^{\infty} \frac{\cancel{(-1)^{n+m}}}{nm} \tilde{f}\left(\frac{n}{a}\right) \tilde{f}\left(\frac{m}{a}\right) \frac{a}{2} \boxed{\delta_{n,m}}\end{aligned}$$

$$\sigma_{\text{rec}}^2 = \frac{a^2}{2\pi^2} \sum_{m=1}^{\infty} \frac{1}{m^2} \tilde{f}^2\left(\frac{m}{a}\right).$$



Another Check

$$\sigma_{\text{rec}}^2 = \frac{a^2}{2\pi^2} \sum_{m=1}^{\infty} \frac{1}{m^2} \tilde{f}^2\left(\frac{m}{a}\right).$$

- For very narrow signals, we have again $\tilde{f}(k) \rightarrow 1$ so that

$$\left(\frac{\sigma_{\text{err}}}{a}\right)^2 = \frac{1}{2\pi^2} \sum_{m=1}^{\infty} \frac{1}{m^2} = \frac{1}{12} \quad \text{as expected....}$$

$$\pi^2/6$$

- This is probably the most complicated way to get the 1/12...



f(x) = Gaussian / Box

- For a Gaussian signal with width σ

$$G(x) = \frac{1}{\sqrt{2\pi}\sigma_s} \exp\left(-\frac{x^2}{2\sigma_s^2}\right) \quad \text{with} \quad \tilde{G}(k) = \exp\left(-2\pi^2 k^2 \sigma_s^2\right)$$

we get

$$\left(\frac{\sigma_{\text{err}}}{a}\right)^2 = \frac{1}{2\pi^2} \sum_{m=1}^{\infty} \frac{\exp\left(-\frac{4m^2\pi^2\sigma_s^2}{a^2}\right)}{m^2}$$

- For a Box of width s a

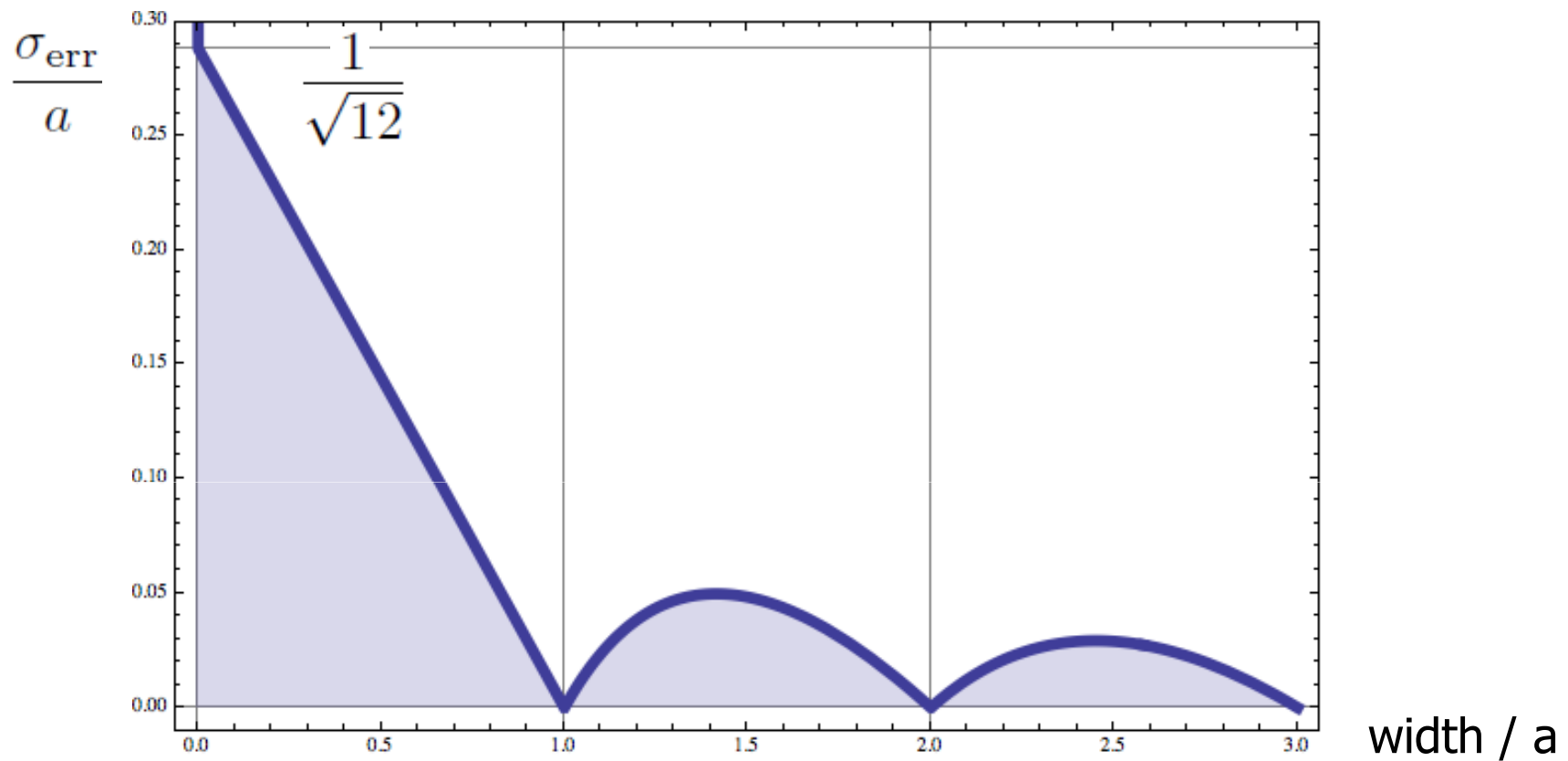
$$\left(\frac{\sigma_{\text{err}}}{a}\right)^2 = \frac{1}{360s^2} - \frac{1}{4\pi^4 s^2} \sum_{m=1}^{\infty} \frac{\cos(2\pi ms)}{m^4}$$

which is zero for integer s thanks to $\Sigma(1/n^4) = \pi^4/90..$



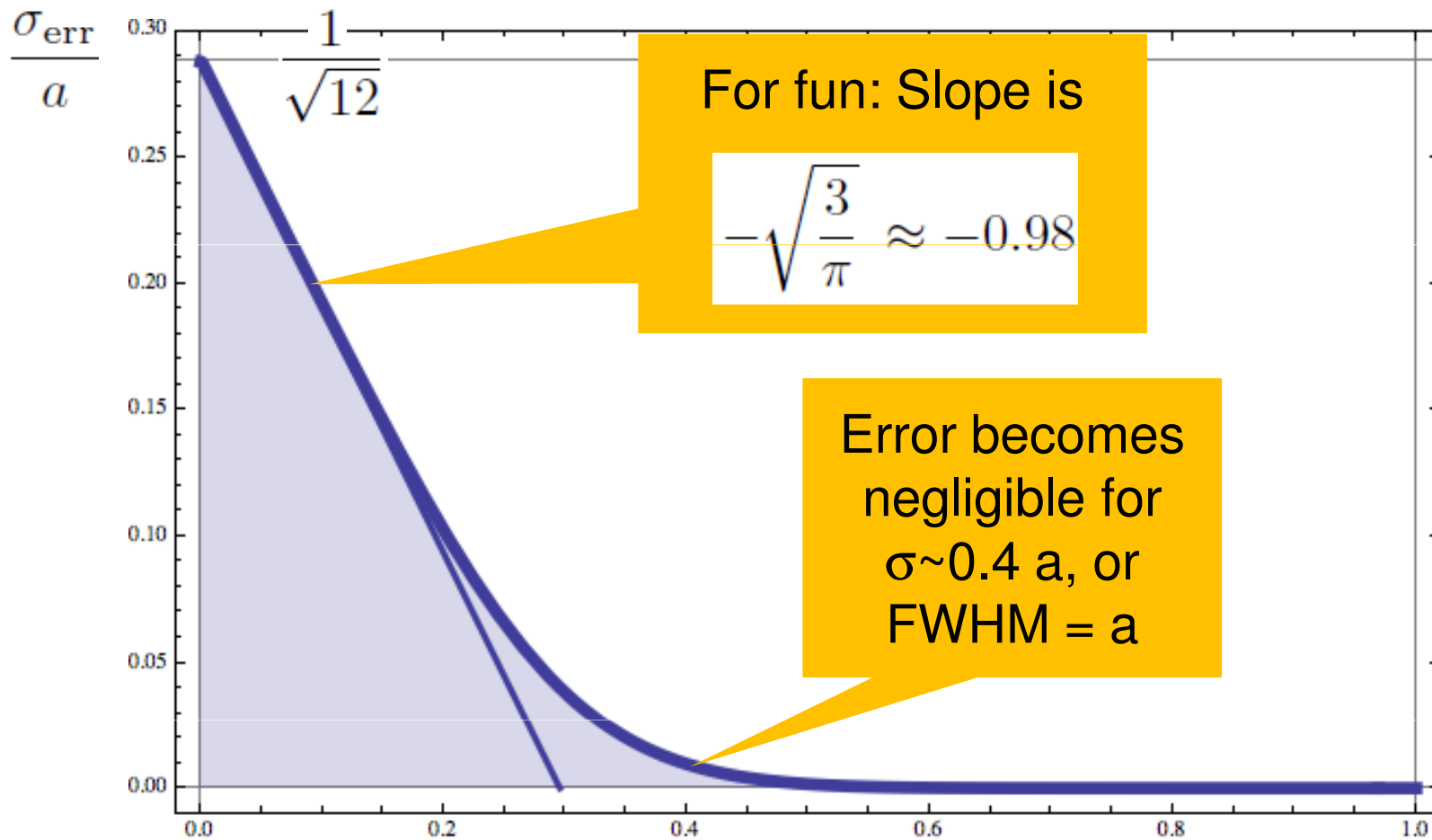
Plot this for $f(x) = \text{Box}$

- Error zero for integer box width.
- Behavior in between not trivial...





Plot this for $f(x) = \text{Gauss}$



The result 'Error ≈ 0 for $\text{FWHM} \approx a$ ' can be found for many pulse shapes. We knew this... but now we know *for sure*...



LIMIT OF SPATIAL RESOLUTION FROM NOISE



The Question

- How is spatial resolution degraded by noise?

- We all know $\sigma_{\text{err}} = \kappa \cdot \sigma_n = \frac{\kappa}{SNR}$

This states, that the resolution degrades with noise ‘linearly to first order’.

- The proportionality κ is empirical. We want to calculate it
- We also want to check what happens with correlated noise
- We want to see what happens to higher order
 - What is this here? It is the distribution of the noise...
- We assume we can reconstruct with CoG (more later...)
- We restrict on a 1D treatment, but 2D is straight forward



Write down \vec{x}_{rek} with noise

- A Signal at \vec{x} is distributed over N strips at positions \vec{x}_i
- Signal on i-th strip is $S_i(\vec{x})$
- The sum of all signals shall be normalized to 1:

$$\sum S_i = 1$$

1

- Assume we can perfectly reconstruct position as center of gravity:

$$\vec{x} = \frac{\sum S_i \vec{x}_i}{\sum S_i} = \sum S_i \vec{x}_i$$

2

- Now assume noise n_i on all strips
- the reconstructed position is:

$$\vec{x}_{\text{rek}}(\vec{x}) = \frac{\sum (S_i + n_i) \vec{x}_i}{\sum (S_i + n_i)} = \frac{\vec{x} + \sum n_i \vec{x}_i}{1 + \sum n_i}$$



Assume noise is small. Get the standard dev.

- This becomes (Taylor Expansion of Denominator):

$$\vec{x}_{\text{rek}}(\vec{x}) = \frac{\vec{x} + \sum n_i \vec{x}_i}{1 + \sum n_i} = \left(\vec{x} + \sum n_i \vec{x}_i \right) \left(1 - \sum n_i + \mathcal{O}(n^2) \right)$$

- The reconstruction *error* is:

$$\vec{x}_{\text{err}}(\vec{x}) = \sum_i n_i (\vec{x}_i - \vec{x}) + \mathcal{O}(n^2).$$

- We need the standard deviation:

$$\sigma_{\text{err}}^2 = \langle \vec{x}_{\text{err}}^2 \rangle - \langle \vec{x}_{\text{err}} \rangle^2$$

Average error is zero!

We need to average over
- **ALL possible positions**
- **ALL noises**



Do the averaging

$$\sigma_{\text{err}}^2 = \langle \vec{x}_{\text{err}}^2 \rangle$$

$$\vec{x}_{\text{err}}(\vec{x}) = \sum_i n_i (\vec{x}_i - \vec{x}) + \mathcal{O}(n^2).$$

$$= \sum_{i,j} \langle n_i n_j \rangle \langle (\vec{x}_i - \vec{x})(\vec{x}_j - \vec{x}) \rangle + \langle \mathcal{O}(n^3) \rangle$$

For **uncorrelated**
noise

$$\langle n_i n_j \rangle = \delta_{ij} \cdot \sigma_n^2$$

$$= \sigma_n^2 \cdot \sum_i \langle (\vec{x}_i - \vec{x})^2 \rangle + \mathcal{O}(\sigma_n^3)$$

- We chose the origin such that

$$\sum_i \vec{x}_i = \vec{0}$$

then this simplifies to:

$$\sigma_{\text{err}}^2 = \sigma_n^2 \left(\sum_{i=1}^N \vec{x}_i^2 + N \langle \vec{x}^2 \rangle \right) + \mathcal{O}(\sigma_n^3).$$

This is κ^2 !

3



Example: Strips

- Consider two strips at $x_1 = -a/2$ and $x_2 = +a/2$ ($N = 2$)
- Signals for a hit at x are

$$S_1(x) = (x_2 - x)/a \text{ and } S_2(x) = (x + x_2)/a$$

- 1, 2 and 3 are fulfilled:

$$S_1 + S_2 = 1; \quad x_1 S_1 + x_2 S_2 = x; \quad x_1 + x_2 = 0$$

- We get $\left(\frac{\sigma_{\text{err}}}{\sigma_n}\right)^2 \approx x_1^2 + x_2^2 + \frac{2}{a} \int_{x_1}^{x_2} x^2 dx = \frac{2}{3} a^2$

- Or

$$\sigma_{\text{err}} = 0.816 \cdot a \cdot \sigma_n$$

- For $\sigma_n = 0.1$ (Signal/Noise = 10), **resolution = 8% · a**
- Resolution is better than optimal binary readout for $S/N > 5.6$



Correlated Noise ?

- For FULLY correlated noise, $n_i = n_j$ and $\langle n_i n_j \rangle = \sigma_n^2$

- We get $\sigma_{\text{err}}^2 \approx \sigma_n^2 N^2 \langle \vec{x}^2 \rangle$

- For the strip example

$$\sigma_{\text{err}} = a \sigma_n / \sqrt{3} = 0.57 a \sigma_n$$

- Correlated noise is less harmful than ‘normal’ noise
- Note: For mixed noise, superimpose both components)
- Note: If the Amplitude of the signal is *KNOWN* (X-ray), noise becomes correlated and resolution improves!



Higher Orders (in noise)

- Noise can have different distributions.

- They have different higher moments:

$$\begin{aligned}\langle n_i^2 \rangle &= \sigma_n^2, \\ \langle n_i^4 \rangle &= \beta \cdot \sigma_n^4.\end{aligned}$$

- They are $\beta := \frac{\int n^4 p(n) dn}{(\int n^2 p(n) dn)^2} = \begin{cases} 3 : \text{Gauss} \\ 9/5 : \text{Box} \\ 1 : \text{Peaks} \end{cases}$

- We need then higher order correlations (not trivial..):

$$\langle n_i \rangle = 0$$

$$\langle n_i n_j \rangle = \delta_{ij} \sigma_n^2$$

$$\langle n_i n_j n_k \rangle = 0$$

$$\langle n_i n_j n_k n_l \rangle = \delta_{ij} \delta_{jk} \delta_{kl} (\beta - 3) \sigma_n^4 + (\delta_{ij} \delta_{kl} + \delta_{ik} \delta_{jl} + \delta_{il} \delta_{jk}) \sigma_n^4$$



Higher Orders

- Repeating the derivation yields

$$\sigma_{\text{err}}^2 = \sigma_n^2 \cdot (A + N B) \cdot \left(1 + 3 \left[\beta - 3 + N \frac{A + 3 N B}{A + N B} \right] \sigma_n^2 \right)$$

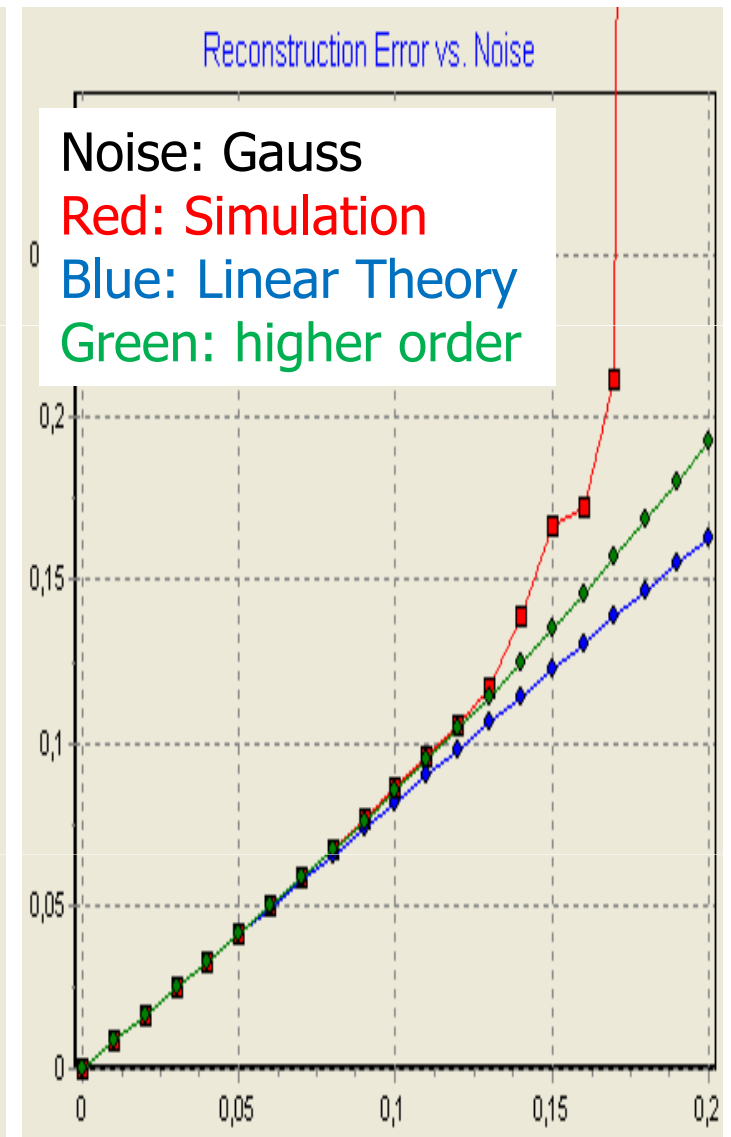
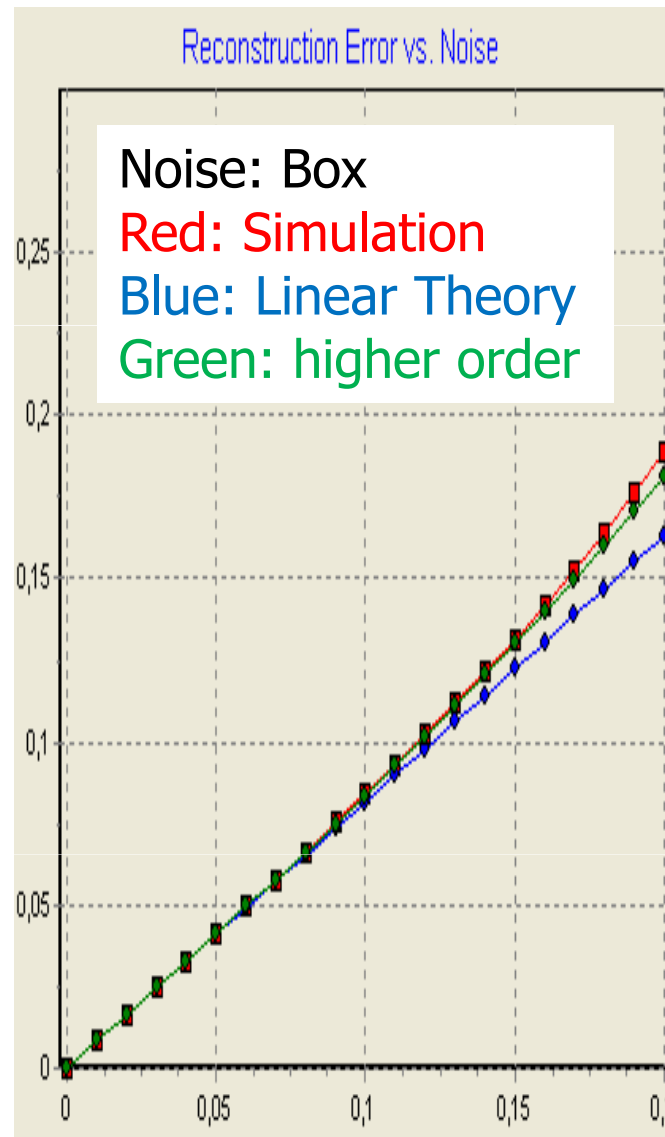
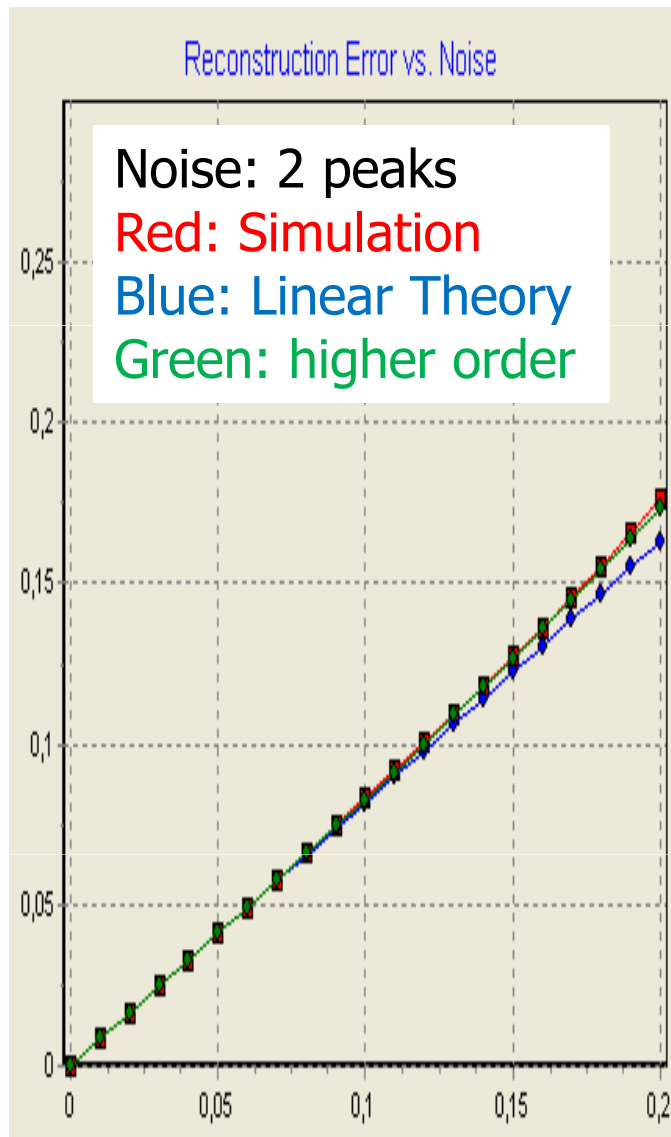
Previous result

Correction

- Only the *correction* depends on the ‘type’ (shape) of noise.
- For small noise, there is *no need* to simulate Gaussian noise:
Randomly adding or subtracting $\pm \sigma_n$ has the same effect!



Is this true? → Small Monte Carlo



- Reconstruction for Gauss noise fails in few cases



2D Structures

- Can be treated similarly
- Observations:
 - Small number of electrodes is good
 - Well confined acceptance 'circle' is good

$$\sigma_{\text{err}}^2 = \sigma_n^2 \left(\sum_{i=1}^N \vec{x}_i^2 + N \langle \vec{x}^2 \rangle \right) + \mathcal{O}(\sigma_n^3).$$

Geometry	σ_{1D} theory		numerical (A=p=1)
	linear	correction	
strips	$\sqrt{\frac{2}{3}} \cdot p$	$\sqrt{1 + 3\beta\sigma_n^2}$	0.8165
square	$\frac{2}{\sqrt{3}} \cdot \sqrt{A}$	$\sqrt{1 + 3(3 + \beta)\sigma_n^2}$	1.1547
hexagon	$\frac{\sqrt{5} \cdot 3^{1/4}}{6} \cdot \sqrt{A}$	$\sqrt{1 + 3\left(\frac{6}{5} + \beta\right)\sigma_n^2}$	0.4905

- Hexagons are best.



BACK To GoG
NOW WITH NOISE



Problems with Centroid

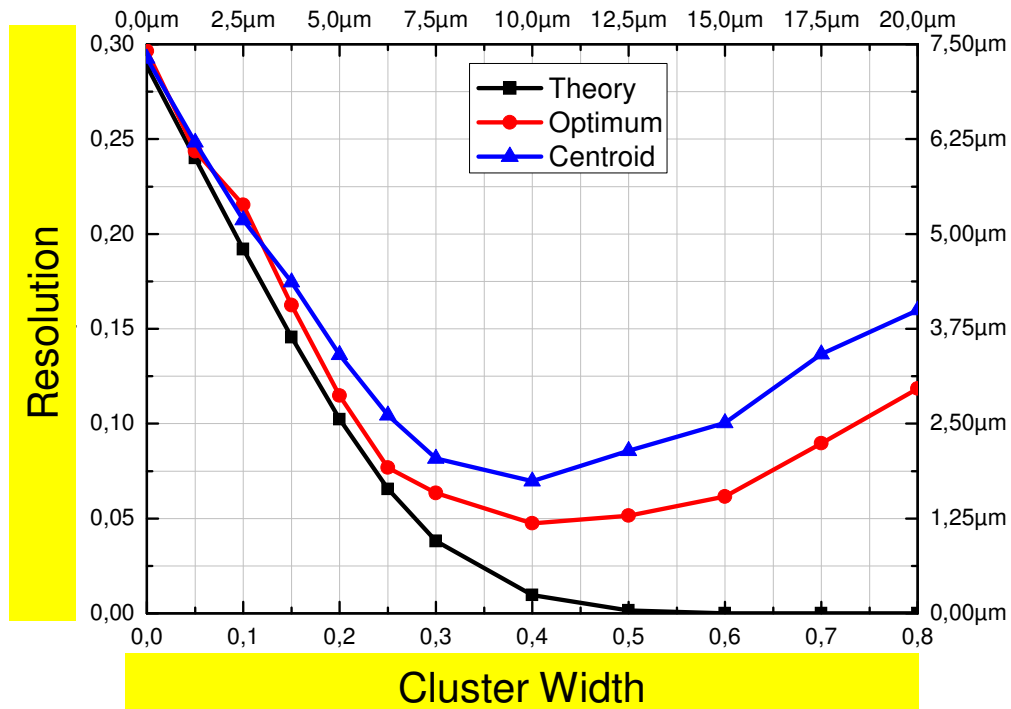
- Resolution for small σ is bad \rightarrow make $f(x)$ wide
- BUT: The 'infinite' sum creates infinite noise (N high)
- Must chose N so that reconstruction is ok.
 - Obviously $N \sim \sigma$
- The choice is fairly arbitrary
- And:
 - In real system, there is often a **threshold** (hits below this are not read out)
 - The reconstructed amplitude is wrong (signals below threshold are lost)
 - Broken pixels need special treatment



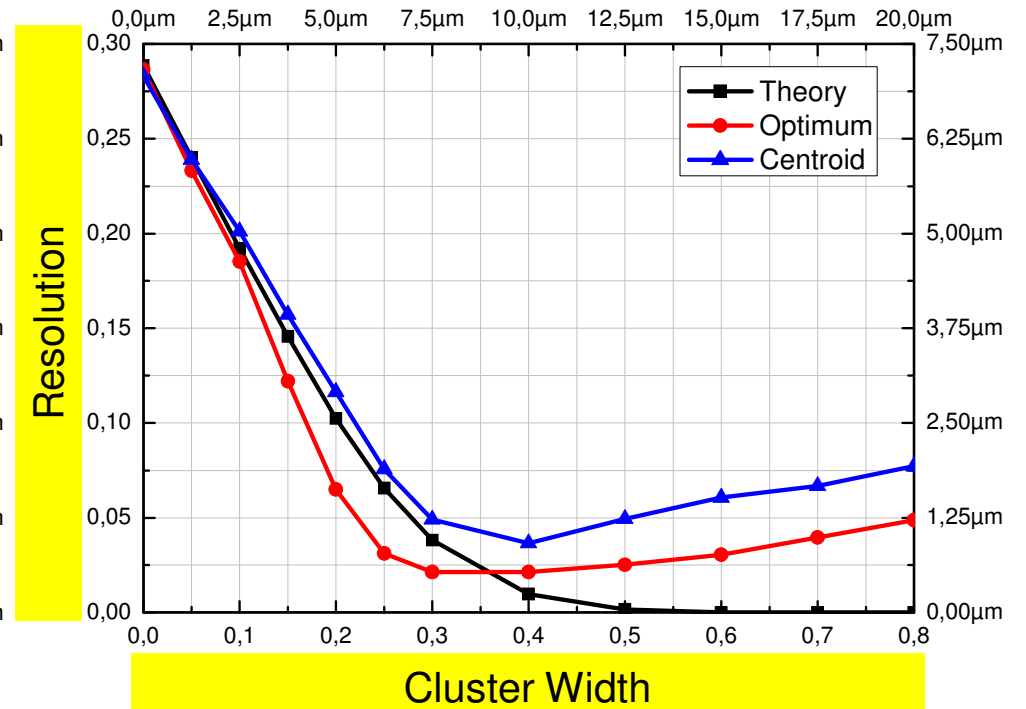
Monte Carlo Simulation

- Many possibilities... I do not go in details

$S/N = 40$



$S/N = 80$



- The optimum signal width is still close to $\text{FWHM} = a$!

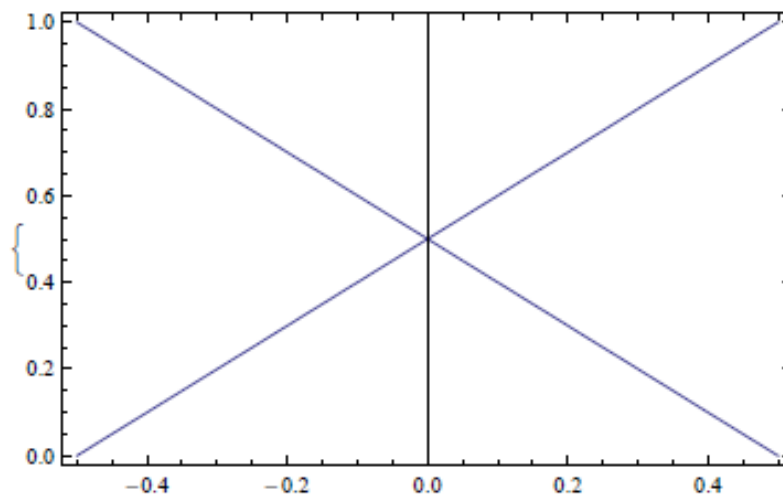


ETA FUNCTION

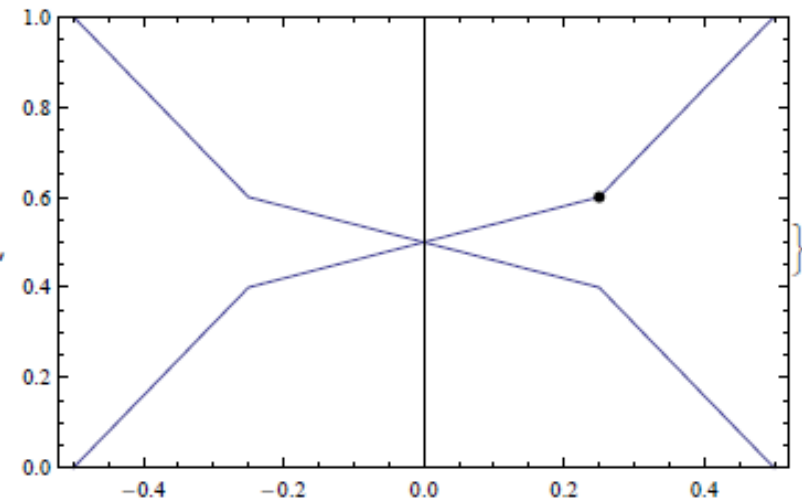


Motivation

- Often the Signals Distribution function (e.g. on 2 strips) is not linear.
- This is related to the ‘famous’ eta-function.



Linear Charge Sharing



Arbitrary function

- The position then cannot be calculated by GoG, but by using the inverse function (or the ‘eta’-lookup table)
- Question: How does resolution depend on $f(x)$?



1D Case: Reconstruction with Inverse Function

- The signals on the two strips shall be

$$S_1(x) = Q f(x)$$

$$S_2(x) = Q - S_1(x) = Q (1 - f(x))$$

(we assume no signal is lost, i.e. we require $S_1 + S_2 = Q$)

- We require

- $f(x)$ is strictly monotonic (obvious)
- $f(x)$ shall be symmetric in x (may not always be the case)

- Obviously

$$x_{rek} = f^{-1} \left[\frac{S_1}{S_1 + S_2} \right]$$



Adding Noise

- With Noise on S_1 and S_2 we get

$$\begin{aligned}
 x_{rek} &= f^{-1} \left[\frac{f(x) + n_1}{1 + n_1 + n_2} \right] && \text{Add noise} \\
 &\approx f^{-1} [(f(x) + n_1)(1 - n_1 - n_2)] && \text{Taylor (as before)} \\
 &\approx f^{-1} [f(x) + n_1(1 - f(x)) - n_2 f(x)] && \text{1st order in noise} \\
 &\approx x + \left. \frac{df^{-1}(s)}{ds} \right|_{f(x)} \cdot [n_1(1 - f(x)) - n_2 f(x)] && \text{Taylor Series for } f^{-1} \text{ around } f(x) \\
 x_{err} &= \frac{n_1(1 - f(x)) - n_2 f(x)}{f'(x)} && \text{A 'forgotten' math theorem: The derivative of the inverse function is the inverse of the derivative}
 \end{aligned}$$



Sigma – Averaging over Noise

- To get

$$\sigma_{err}^2 = \langle x_{err}^2 \rangle - \langle x_{err} \rangle^2 \quad x_{err} = \frac{n_1(1 - f(x)) - n_2 f(x)}{f'(x)}$$

- we average first over noise. We get

$$\sigma_{err}^2 = \sigma_n^2 \left\langle \frac{1 - 2f + 2f^2}{f'^2} \right\rangle + 2 \langle n_1 n_2 \rangle \left\langle \frac{f^2 - f}{f'^2} \right\rangle$$

- Coefficients depend on the signal shape
- They are small where the response function is steep (obvious..)
- For uncorrelated noise, only the first term matters

$$\frac{\sigma_{err}^2}{\sigma_n^2} = \left\langle \frac{1 - 2f + 2f^2}{f'^2} \right\rangle$$

Average over
position



Back to linear Interpolation

- What does this mean for linear interpolation, $f(x) = x+0.5$?
- Let us first look at the *position dependent* error

$$\frac{\sigma_{err}^2}{\sigma_n^2} = \frac{1 - 2f + 2f^2}{f'^2} = \frac{1}{2} + 2x^2$$

- This is NOT constant. It doubles at the edges !!!
- The average error is

$$\frac{\sigma_{err}^2}{\sigma_n^2} = \int_{-1/2}^{1/2} \left(\frac{1}{2} + 2x^2 \right) = \frac{2}{3} \quad (\text{s. page 28})$$

as before.



Finding New Distribution functions

- Very exciting: Can we find a $f(x)$ such that the integral is **better** than with linear interpolation
 - Probably not. But let's see...
- Can we find a distribution function so that the error is independent of position?
- One line of Mathematica is enough:

$$f_{flat}(x) = \frac{1}{2} (1 - \text{Sinh}[2x \text{ArcSinh}(s)])$$

- The average σ^2 is 0.643, which is (a bit) **better** than $2/3$!!
- We found a distribution which is better than lin. interpolation!



Summary: What did we learn ?

- Basic Algebra is fun.....
- CoG is 'perfect' when signal width = strip width
- Larger signals are worse when there is noise
- Ideal κ for strips is 0.816
- Analogue readout for $S/N < 6$ is useless.
- Noise shape does not matter for $S/N > 10$
- Correlated noise is less harmful
- Hexagons have better resolution
- Linear interpolation has more error at the edges.
- There is a better reconstruction function than linear
 - but the difference is negligible....



Thank you for your attention!